Reg. No.



MANIPAL INSTITUTE OF TECHNOLOGY MANIPAL tion of Manipal University

SEVENTH SEMESTER B.TECH. (INSTRUMENTATION AND CONTROL ENGG.) END SEMESTER EXAMINATIONS, NOV/DEC 2016

SUBJECT: ROBUST CONTROL [ICE 439]

Time: 3 Hours

MAX, MARKS: 50

Instructions to Candidates:

- ✤ Answer ANY FIVE FULL questions.
- ✤ Missing data may be suitably assumed.
- A plant has state space description $\dot{x} = \begin{bmatrix} -0.2 & -1 \\ 1 & 0 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u, y = \begin{bmatrix} 0 & 1 \end{bmatrix} x$. Find 5 1A. the corresponding transfer function G(s) and obtain $\| \|_{\infty}$ of G(s). Also, suggest a signal u(t) for which $\|y\|_2 = \|G\|_{\infty} \|u\|_2$.
- A system with transfer function $G(s) = \frac{1}{10s+1}$ has a disturbance input d(t) known to 1B. 3 have energy bound $\|d\|_2 \leq 0.4$. Find the best estimate of the ∞ - norm of the output y(t).
- Compute both $\|\mathbf{y}\|_2$ and $\|\mathbf{y}\|_{\infty}$ for a signal $\mathbf{y}(\mathbf{t}) = \mathbf{e}^{-\mathbf{t}}$, if \mathbf{y} is defined for the values 1C. 2 of time $\mathbf{t} \in [\mathbf{0} \infty)$.
- Consider the standard feedback loop shown in Fig. Q2A where $P(s) = \frac{1}{s+1}$, C(s) = K and F(s) = 1. Find the controller gain values K > 0 such 2A. 5 that the steady-state absolute error |e(t)| is less than or equal to 0.01 when inputs r(t) are unit step and $\sin \omega t$, $0 \le \omega \le 4$ respectively with n = d = 0.
- 2B. Assume that unity feedback system as shown in Fig. Q2A is internally stable 3 and n = d = 0. Show that if input r(t) is a ramp then $e(t) \to 0$ as $t \to \infty$ if and
- only if sensitivity transfer function S(s) has at least two zero at the origin. Consider two plants $P_1(s) = \frac{1}{s-1}$, $P_2(s) = \frac{1}{10s-1}$. Do you think one is harder to 2C. 2 stabilize than the other? Illustrate your answer.
- 3A. Derive robust performance condition for the feedback control system having 5 multiplicative uncertainty plant model.
- Consider the system $G(s) = \begin{bmatrix} \frac{1}{s+c} & \frac{4}{s+8} \end{bmatrix}$, $c \in R$ and the static controller $K = \begin{bmatrix} -4 \\ 1 \end{bmatrix}$. 3B. 3 What conclusion about the nominal stability of the closed loop system be deduced with the small gain theorem using ∞ - norm of G(s) for $c = \{-1, 1, 3\}$?
- 3C. If $W_{P}(s)$ is the weighting function and L(s) is the loop transfer function for the unity feedback control system, show that $\|W_p S\|_{\infty} < 1 \Leftrightarrow |W_p(j\omega)| < |1 + L(j\omega)|, \forall \omega.$
- Prove that the system with transfer function $P(s) = \frac{1}{s+0.5+e^{-\tau s}}$ is stable for any 5 4A. $\tau \in [0, 1).$

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- A plant has transfer function $G(s) = e^{-\tau s} \frac{1}{s^2 + 0.4s + 1}$. The delay τ is unknown, but lies 4B. 3 between 0 and 4. A proportional feedback controller is implemented with gain K. Use small gain theorem to determine the maximum gain K that preserves the stability for all possible values of τ and interpret your answer in terms of the Nyquist criterion. 2
- 4C. Illustrate parametric uncertainty in plant model with example.
- Find an internally stabilizing controller C for the plant model $P(s) = \frac{3}{s-4}$ of unity 5 5A. feedback system as shown in Fig. Q2A so that tracking error e goes to 0, when r is a ramp and n = d = 0.
- 5B. Derive robust stability condition for the feedback system shown in Fig. **Q5B**.

5C. Show that
$$|||W_PS| + |W_{\Delta}T|||_{\infty} < 1 \Rightarrow \max_{\omega} \overline{\sigma} \begin{bmatrix} W_PS \\ W_{\Delta}T \end{bmatrix} < \frac{1}{2}$$

- Consider $P(s) = \frac{1}{s}$ and $W_p(s) = \frac{100}{s+1}$. Perturb P to $P(s) = 1/(s+\epsilon), \epsilon > 0$. Find 6A. 5 the controller C (internally stabilizing) so that $\|W_P S\|_{\infty} < 1$. Does the resulting controller C solve the performance design problem for the original P? Again, factor **P** as $P = P_1 P_2$, $P_1 = \frac{1}{s+1}$, $P_2 = \frac{s+1}{s}$. Solve the performance design problem for **P**₁; let C₁ be the solution. Set $C = C_1/P_2$. Does the resulting C solve the performance design problem for the original **P**? If so, explain why.
- Show that set of all controllers for which the feedback system shown in Fig. Q2A is **6B** 3 internally stable is given by $C = \left\{ \frac{X + MQ}{Y - NQ} : Q \in R\mathcal{H}_{\infty} \right\}$ A plant with transfer function $G = \frac{1}{0.11s^2 + 1.2s + 1}$ is to be modelled with the nominal
- 6C 2 transfer function $G_n = \frac{1}{s+1}$. Suggest a suitable first order weighting function W(s) such that $G(s) = G_n(s) + \Delta W(s)$ for some Δ satisfying $\|\Delta\|_{\infty} \leq 1$



Fig. O5B

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