

# ECE 3152 Linear Control Theory

## MATHEMATICAL MODELS OF SYSTEMS

### Electrical Network Transfer Functions

Table: Voltage-current, voltage-charge, and impedance relationships for capacitors, resistors, and inductors

Component	Voltage-current	Current-voltage	Voltage-charge	Impedance $Z(s) = \frac{V(s)}{I(s)}$
Capacitor	$v(t) = \frac{1}{C} \int_0^t i(\tau) d\tau$	$i(t) = C \frac{dv(t)}{dt}$	$v(t) = \frac{1}{C} q(t)$	$\frac{1}{Cs}$
Resistor	$v(t) = Ri(t)$	$i(t) = \frac{1}{R} v(t)$	$v(t) = R \frac{dq(t)}{dt}$	$R$
Inductor	$v(t) = L \frac{di(t)}{dt}$	$i(t) = \frac{1}{L} \int_0^t v(\tau) d\tau$	$v(t) = L \frac{d^2q(t)}{dt^2}$	$Ls$

Note: The following set of symbols and units is used:  $v(t)$  - V (volts);  $i(t)$  - A (amps);  $q(t)$  - Q (coulombs);  $C$  - F (farads),  $R$  -  $\Omega$  (ohms);  $L$  - h (henries)

### Translational Mechanical System Transfer functions

Table 2.4: Force-velocity, force-displacement, and impedance translational relationships for springs, viscous damper, and mass.

Component	Force-velocity	Force-displacement	Impedance $Z_m(s) = \frac{F(s)}{X(s)}$
Spring	$f(t) = K \int_0^t v(\tau) d\tau$	$f(t) = Kx(t)$	$K$
Viscous damper	$f(t) = Bv(t)$	$f(t) = B \frac{dx(t)}{dt}$	$Bs$
Mass	$f(t) = M \frac{dv(t)}{dt}$	$f(t) = M \frac{d^2x(t)}{dt^2}$	$Ms^2$

Note: The following set of symbols and units is used:  $f(t)$  - N (newtons);  $x(t)$  - m (meters);  $v(t)$  - m/s (meters/second);  $K$  - N/m (newton/meter),  $B$  - N-s/m (newton-seconds/meter);  $M$  - kg (kilograms=newton-second<sup>2</sup>/meter)

### Rotational Mechanical System Transfer functions

Table 2.5: Torque-angular velocity, torque angular displacement, and impedance rotational relationships for springs, viscous damper, and inertia.

Component	Torque-angular velocity	Torque-angular displacement	Impedance $Z_m(s) = \frac{T(s)}{\theta(s)}$
Spring	$T(t) = K \int_0^t \omega(\tau) d\tau$	$T(t) = K\theta(t)$	$K$
Viscous damper	$T(t) = B\omega(t)$	$T(t) = B \frac{d\theta(t)}{dt}$	$Bs$
Inertia	$T(t) = J \frac{d\omega(t)}{dt}$	$T(t) = J \frac{d^2\theta(t)}{dt^2}$	$Js^2$

Note: The following set of symbols and units is used:  $T(t)$  – N-m (newton-meters);  $\theta(t)$  – rad (radians);  $\omega(t)$  – rad/s (radians/second);  $K$  – N-m/rad (newton-meter/second),  $B$  – N-m-s/rad (newton-meter-second/radians);  $J$  – kg-m<sup>2</sup> (kilogram-meters<sup>2</sup>=newton-meter-second<sup>2</sup>/radian)

### Transfer Functions for Systems with Gears

$T_1(t)$ - input torque

$\theta_1(t)$  – angle of rotation of input gear

$r_1$  – radius of input gear

$N_1$  – number of teeth on input gear

$T_2(t)$ - output torque

$\theta_2(t)$  – angle of rotation of output gear

$r_2$  – radius of output gear

$N_2$  – number of teeth on output gear

$$a) \frac{\theta_2}{\theta_1} = \frac{r_1}{r_2} = \frac{N_1}{N_2}$$

$$b) T_1\theta_1 = T_2\theta_2$$

$$c) \frac{T_2}{T_1} = \frac{\theta_1}{\theta_2} = \frac{N_2}{N_1}$$

Note: Rotational mechanical impedances can be reflected through gear trains by multiplying the mechanical impedance by the ratio

$$\left[ \frac{\text{Number of teeth of gear on destination shaft}}{\text{Number of teeth of gear on source shaft}} \right]^2$$

### TRANSIENT RESPONSE SPECIFICATIONS

#### Second order system

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

System is un-damped when  $\xi = 0$

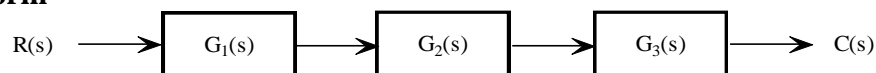
System is under damped when  $0 < \xi < 1$

System is critically damped when  $\xi = 1$

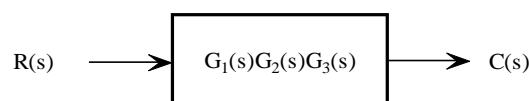
System is over damped when  $\xi > 1$

### REDUCTION OF MULTIPLE SUBSYSTEMS

#### Cascade form

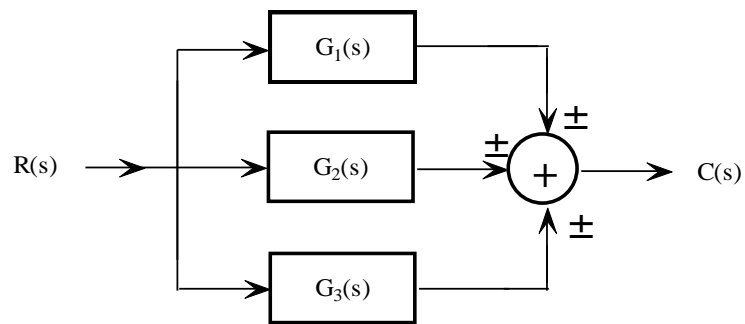


(a) Cascaded subsystem

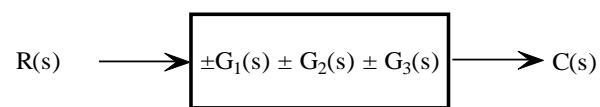


(b) Equivalent transfer function

## Parallel form

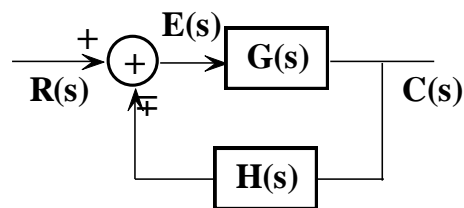


(a) Parallel subsystem

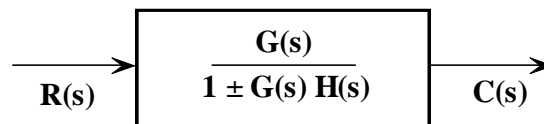


(b) Equivalent transfer function

## Feedback Form



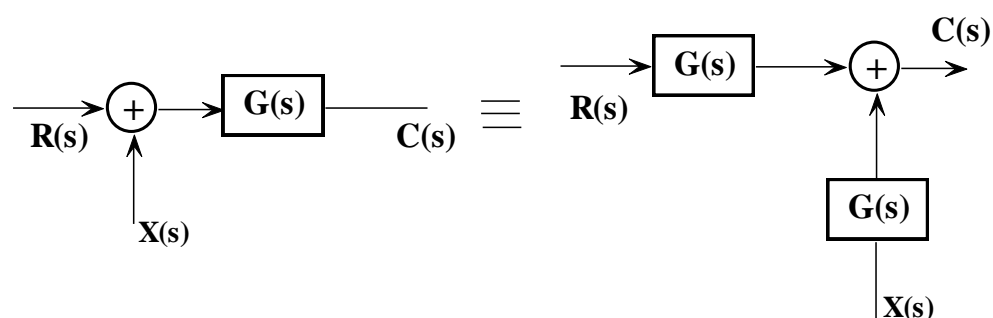
Feedback control system



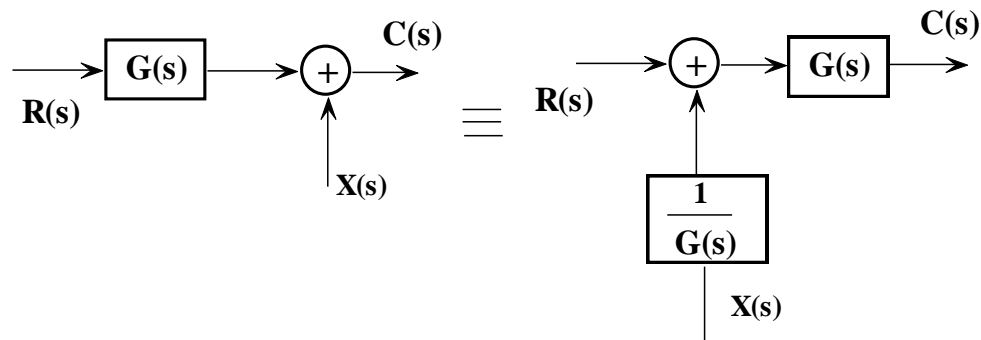
Equivalent transfer function

## Moving blocks to create familiar forms

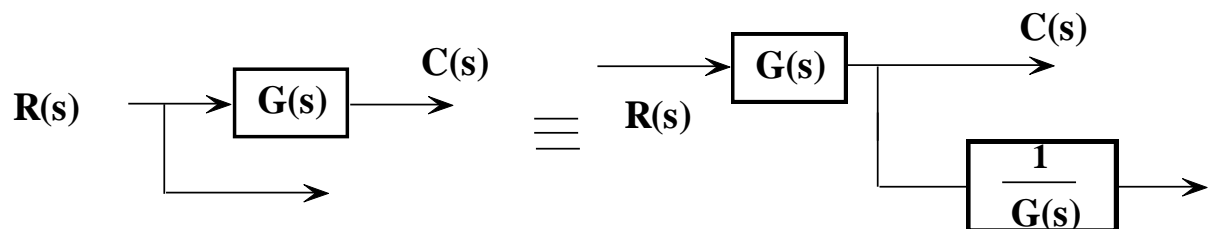
### i) Moving a block to the left past a summing junction



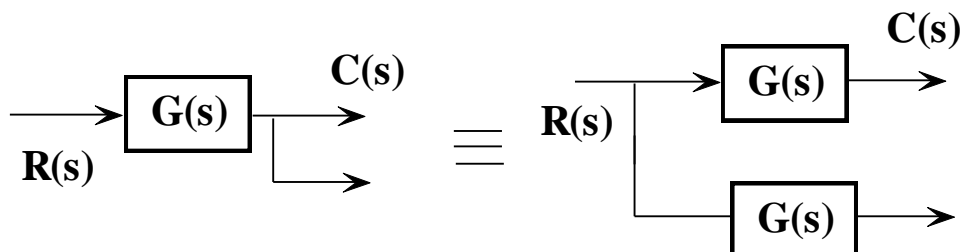
**ii) Moving a block to the right past a summing junction**



**iii) Moving a block to the left past a pickoff point**



**iv) Moving a block to the right past a pickoff point**



**SIGNAL-FLOW GRAPHS**

**Mason's rule**

The transfer function  $\frac{C(s)}{R(s)}$  of a system represented by a signal-flow graph is

$$\frac{C(s)}{R(s)} = \frac{\sum_k T_k \Delta_k}{\Delta}$$

Where

$k$  = number of forward paths

$T_k$  = the  $k^{\text{th}}$  forward-path gain

$\Delta = 1 - \sum \text{loop gains} + \sum \text{product of non-touching loop gains taken two at a time} - \sum \text{product of non-touching loop gains taken three at a time} + \sum \text{product of non-touching loop gains taken four at a time} \dots\dots\dots$

$\Delta_k = \Delta - \sum \text{loop gain terms in } \Delta \text{ that touch the } k^{\text{th}} \text{ forward path. } \Delta_k \text{ is formed by eliminating from } \Delta \text{ those loop gains that touch the } k^{\text{th}} \text{ forward path.}$

## STABILITY

### Routh-Hurwitz criterion

Considering the characteristic equation

$$a_0 s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{n-1} s + a_n = 0$$

Form the determinants from the coefficients of the characteristic equation. There are  $n$  determinants for  $n^{\text{th}}$  order system.

The system is stable if and only if the value of each determinant is greater than zero.

## STEADY-STATE RESPONSE SPECIFICATIONS

<i>Input</i>	<i>Static error constants</i>	<i>Steady state errors</i>
<i>Step function</i>	$K_P = \lim_{s \rightarrow 0} G(s)$	$\frac{1}{1+k_p}$
<i>Ramp function (t)</i>	$K_r = \lim_{s \rightarrow 0} S G(s)$	$\frac{1}{k_v}$
<i>Parabolic function (t<sup>2</sup>)</i>	$K_a = \lim_{s \rightarrow 0} S^2 G(s)$	$\frac{1}{k_a}$

## FREQUENCY RESPONSE SPECIFICATIONS

- a) Resonant peak  $M_r = \frac{1}{2\xi\sqrt{1-\xi^2}}$  ;  $\xi = 0.707$
- b) Resonant frequency  $\omega_r = \omega_n \sqrt{1-2\xi^2}$  ;  $\xi < 0.707$
- c) Bandwidth
$$BW = \omega_n \sqrt{(1-2\xi^2) + \sqrt{4\xi^4 - 4\xi^2 + 2}}$$
- d) Gain margin  $= \frac{1}{|G(j\omega_p)H(j\omega_p)|}$ ;  
Gain margin in dB = 0 – gain in dB at  $\omega = \omega_p$   
 $\omega_p$  is phase crossover frequency
- e) Phase margin =  $180^\circ + \text{phase at } \omega = \omega_g$   
 $\omega_g$  is gain crossover frequency

## ROOT LOCUS TECHNIQUES

- a) Open-loop transfer function  $G(s)H(s) = K \frac{N(s)}{D(s)}$
- b) Magnitude criterion  $|G(j\omega)H(j\omega)| = 1$
- c) Angle criterion  $\angle G(j\omega)H(j\omega) = \pm 180^\circ$
- d) For breakaway and break-in points solve  $\frac{dK}{ds} = 0$
- e) Centroid of asymptote  $\sigma = \frac{\sum \text{open loop poles} - \sum \text{open loop zeros}}{n-m}$   
 $n$ = number of open loop poles;  $m$ = number of open loop zeros
- f) Angle of intercept of asymptotes  $\beta = \left(\frac{2l+1}{n-m}\right) 180^\circ$ ;  $l = 0, 1, 2, \dots, (n-m-1)$
- g) Angle of departure from the complex pole  $s = -a + jb$   
 $\phi = 180^\circ + \text{Angle of } \{(s+a-jb)G(s)H(s)\} \text{ at } s = -a + jb$   
Angle of departure from the complex pole  $s = -a - jb$   
 $\phi = 180^\circ + \text{Angle of } \{(s+a+jb)G(s)H(s)\} \text{ at } s = -a - jb$

- h) Angle of arrival at the complex zero  $s = -a + jb$   
 $\phi = 180^\circ - \text{Angle of } \left\{ \frac{G(s)H(s)}{(s+a-jb)} \right\} \text{ at } s = -a + jb$   
 Angle of arrival at the complex zero  $s = -a - jb$   
 $\phi = 180^\circ - \text{Angle of } \left\{ \frac{G(s)H(s)}{(s+a+jb)} \right\} \text{ at } s = -a - jb$
- i) Value of K at the point  $s=s_o$  on the root locus  $K = \left| \frac{D(s_o)}{N(s_o)} \right|$
- j) To find the  $j\omega$  axis crossing, use Routh Hurwitz criterion.

## STATE-SPACE REPRESENTATION

- a) State Equation:  $\dot{X} = AX + Br$   
 b) Output equation:  $y = CX + Dr$   
 c) Eigen Values: Solve  $|\lambda I - A| = 0$   
 d) Transfer function:  $T(s) = \frac{Y(s)}{R(s)} = C [sI - A]^{-1} B + D$   
 e) State transition matrix:  $e^{At} = L^{-1} [sI - A]^{-1}$   
 f) Zero input response:  $Y_{ZIR}(s) = C [sI - A]^{-1} X(0)$   
 g) Zero state response:  $Y_{ZSR}(s) = C [sI - A]^{-1} B R(s)$

**Gain margin** =  $20 \log \frac{1}{|\alpha|}$  ; where  $\alpha$  is the magnitude of the function at  $\phi = -180^\circ$ .

**Phase Margin** =  $\angle GH(j\omega_1) + 180^\circ$ ; Where  $\omega_1$  is the gain cross over frequency.

**Lag compensator**  $G_c(s) = Kc \frac{s+1/T}{s+1/\alpha T}$  ;  $\alpha > 1$

**Lead compensator**  $G_c(s) = Kc \frac{s+1/T}{s+1/\alpha T}$  ;  $\alpha < 1$

**PID Controller**  $G_c(s) = K_p + \frac{K_i}{s} + K_d s$

## Laplace Transforms

Time domain	Laplace Domain
$\delta(t) = \text{unit Impulse}$	1
$u(t) = \text{unit step}$	$\frac{1}{s}$
$t = \text{ramp}$	$\frac{1}{s^2}$
$e^{-at}$	$\frac{1}{s+a}$
$te^{-at}$	$\frac{1}{(s+a)^2}$
$\frac{1}{a}(1 - e^{-at})$	$\frac{1}{s(s+a)}$
$e^{-at} - e^{-bt}$	$\frac{b-a}{(s+a)(s+b)}$
$\sin(bt)$	$\frac{b}{s^2 + b^2}$
$\cos(bt)$	$\frac{s}{s^2 + b^2}$

$e^{-at} \sin(bt)$	$\frac{b}{(s+a)^2 + b^2}$
$e^{-at} \cos(bt)$	$\frac{s+a}{(s+a)^2 + b^2}$
$\text{Sinh}(bt)$	$\frac{b}{s^2 - b^2}$
$\text{Cosh}(bt)$	$\frac{s}{s^2 - b^2}$
$\delta(t - kT)$	$e^{-kTs}$
$t^2$	$\frac{2}{s^3}$
Initial value Theorem	$\lim_{t \rightarrow 0} f(t) = \lim_{s \rightarrow \infty} sf(s)$
Final value Theorem	$\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sf(s)$
First differentiation $f'(t)$	$sF(s) - f(0)$
Second differential $f''(t)$	$s^2F(s) - sf(0) - f'(0)$