Reg. No.



## MANIPAL INSTITUTE OF TECHNOLOGY

(A constituent unit of MAHE, Manipal)

## FIRST SEMESTER M.TECH. (POWER ELECTRONICS & DRIVES / ELECTRIC VEHICLE TECHNOLOGY) END SEMESTER EXAMINATIONS, JANUARY 2023

## **DESIGN OF CONTROL SYSTEMS [ELE 5152]**

**REVISED CREDIT SYSTEM** 

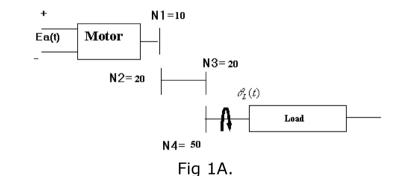
ïme: 3 Hours	Date: 03 January 2023	Max. Marks: 50
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Time: 3 Hours Instructions to Candidates:

- ✤ Answer ALL the questions.
- Missing data may be suitably assumed.

**1A.** Derive the transfer function  $G(s) = \frac{\theta_L(s)}{E_a(s)}$  of this motor, if it drives an

inertia load of 100 kg-m<sup>2</sup> through a gear train as shown in Fig 1. If the inertia and damping of the armature are 5kg-m<sup>2</sup> and 1 N-m/rad respectively. The torque – speed relation is given by  $T_m = -0.1 w_m + 100$ 



**1B.** An industrial Robot system with controller is represented by the closed loop transfer function

$$G(s) = \frac{K}{(s+10)(s^2+4s+10)}$$

- i) Check for second order approximation
- ii) Obtain the time domain specifications percentage overshoot, settling time and peak time of the step response. (04)
- **1C.** Realize the PI controller using an operational amplifier circuit. **(02)**
- **2A.** For the transfer function  $\frac{Y(s)}{R(s)} = \frac{(s+4)}{(s+1)(s+2)(s+5)}$ , derive the state model in i) controllable canonical form ii) parallel form. Using parallel form analyze the controllability and observability of the system. **(04)**

(04)

The state space model of DC -DC Converter is given as 2B.  $\begin{bmatrix} i_L \\ v_C \end{bmatrix} = \begin{bmatrix} -3 & 1 \\ 0 & -2 \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix} + \begin{bmatrix} 1 \\ 2 \end{bmatrix} u; y = \begin{bmatrix} 2 & 1 \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix}$ , i) design a state feedback controller that places the closed loop poles at -3 + i4. Use coefficient matching method. (03) The state space model of DC -DC Converter is given as **2C**.  $\begin{bmatrix} i_L \\ v_C \end{bmatrix} = \begin{bmatrix} -3 & 1 \\ 0 & -2 \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix} + \begin{bmatrix} 1 \\ 2 \end{bmatrix} u; y = \begin{bmatrix} 2 & 1 \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix}$ , i) design an observer 5 times faster than the controller loop poles at  $-3 \pm i4$ . Use coefficient matching method. (03) A Linear Time Invariant system is represented by the state 3A. equation  $\dot{x} = \begin{bmatrix} -2 & 1 \\ 0 & -2 \end{bmatrix} x$ , Using Lyapunov stability criterion, assess the stability of the system, and also find the corresponding Lyapunov function. (04) 3B. Derive state regulator using Lyapunov Method and optimal control theory. (03) Discuss the significance of controllability and observability in 3C. controller design? (03) 4A. For a system represented by a state model  $\dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$ ,  $x(0) = \begin{bmatrix} 1 & -1 \end{bmatrix}^T$ , design using reduced matrix Riccati equation, an optimal control law that minimizes the performance index  $J = \int_0^\infty (4x_1^2 + u^2) dt$ . Comment on stability of the compensated (04) system. Discuss the procedure for designing a stable PID controller using **4B**. Zeigler Nichols tuning method. (03) Discuss the Procedure for LAG compensator design using frequency **4C**. domain methods (03) Explain System identification with an example. (02) 5A. 5B. Derive Model reference control with suitable example (03) Design a sliding mode control scheme for a single mass motion 5C. control. Show how can we extend this for a Buck converter application. (05)